

SUB-MICRON FBAR MODELING INCLUDING FEM SIMULATIONS

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October 2010



Outline

- Motivation
- Introduction
- FBAR Fabrication Using Bulk Micromachining
- Analytical Modeling
- FEM modeling
- Conclusions

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Motivation

Applications:

- wireless communication systems as mobile phones, wireless local area networks (WLAN)
- navigation
- various applications involving data transmission.



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Introduction

- Electricity : electrical displacement

$$\mathbf{D} = \epsilon_T \mathbf{E}$$

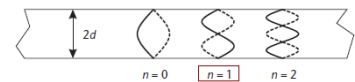
- Solid mechanics : Hooke's law

$$\mathbf{S} = s_E \mathbf{T}$$

- Piezoelectric material : both laws are coupled!

$$\mathbf{S} = s_E \mathbf{T} + d^T \mathbf{E}$$

$$\mathbf{D} = d \mathbf{T} + \epsilon_T \mathbf{E}$$



Mechanical resonances in a plate of thickness $2d$

(RF Bulk Acoustic Wave Filters for Communications, Ken-ya Hashimoto
Editor, ARTECH HOUSE, 2009)

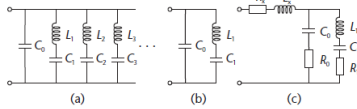
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Introduction

Butterworth-Van Dyke circuit:



(a) Multiresonant BVD circuit
(b) Single resonant circuit
(c) Circuit taking into account losses

(RF Bulk Acoustic Wave Filters for Communications,
Ken-ya Hashimoto Editor, ARTECH HOUSE, 2009)

$$Z(\omega) = \frac{V}{I} = \frac{1}{j\omega C_0} \left[1 - K^2 - \sum_n \frac{\omega^2 k_n^2}{\omega_{a,n}^2 - \omega^2} \right]$$

$$k_n^2 = \frac{8K^2}{[(2n+1)\pi]^2} \rightarrow (b) \quad Z(\omega) = \frac{j(\omega L_1 - 1/\omega C_1)}{1 - \omega^2 C_0 L_1 + C_0 / C_1}$$

$$\omega_r = \frac{1}{\sqrt{L_1 C_1}} \quad \omega_a = \omega_r \cdot \sqrt{1 + \frac{C_1}{C_0}}$$

$$(c) \quad Z(\omega) = j\omega L_s + R_s + \left[\frac{1}{R_0 + 1/j\omega C_0} + \frac{1}{R_1 + j(\omega L_1 - 1/\omega C_1)} \right]^{-1}$$

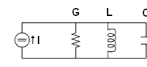
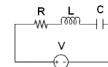
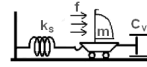
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Introduction

Correspondences between mechanical and electric quantities:



$f = m \frac{d^2 u}{dt^2} + c_v \dot{u} + k_s \int \dot{u} dt$
 f , exerting force (N)
 m , mass (kg)
 c_v , viscous coeff. (Nm/s)
 $\dot{u}$ velocity (m/s)
 $(k_s)^{-1}$ elastic compliance (m/N)
 $\int \dot{u} dt$, force (N)
 $1/2 m \dot{u}^2$, kinetic energy (kgm²/s²)
 $1/2 k_s u^2$, elastic energy (kgm²/s²)
 $\int \dot{u} dt$, power (W=Nm/s)
 $P = \dot{U}$, complex power (W=Nm/s)
 P/U , impedance (W=Nm/s)

$V = L \frac{di}{dt} + R i + \frac{1}{C} \int i dt$
 V , voltage generator (V)
 L , inductance (H)
 R resistance (Ω)
 C capacitance (F)
 i current (A)
 $\int i dt$, voltage (V)
 v , voltage (V)
 $1/2 L i^2$, magnetic energy (HA²)
 $1/2 C v^2$, electrostatic energy (FV²)
 $v \cdot i$, power (W=VA)
 $V \cdot I$, complex power (W=VA)
 V/I , impedance (Ω=V/A)

$I = C \frac{dv}{dt} + G v + \frac{1}{L} \int v dt$
 I , current generator (A)
 C capacitance (F)
 G conductance (S)
 L , inductance (H)
 v , voltage (V)
 i current (A)
 $1/2 C v^2$, electrostatic energy (FV²)
 $1/2 L i^2$, magnetic energy (HA²)
 $i \cdot v$, power (W=AV)
 $I \cdot V$, complex power (W=AV)
 I/V , admittance (S=A/V)

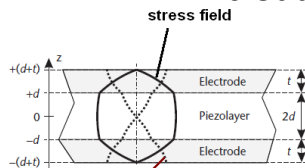
(Modeling and characterization of piezoelectric transducers by means of scattering parameters, Part I: Theory,
Rudi P. Paganelli, A. Romani, A. Golfarelli, M. Magi, E. Sangiorgi, M. Tartagni, Sensors and Actuators A: Physical, 2010)

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Introduction

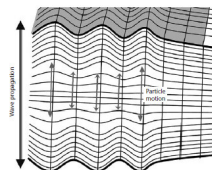


Geometry of a resonator with electrodes in the
case when $Z_p > Z_e$

(RF Bulk Acoustic Wave Filters for Communications,
Ken-ya Hashimoto Editor, ARTECH HOUSE, 2009)

The displacement:

$$u(z) = \begin{cases} a \sin(k_p z) & -d \leq z \leq d \\ b \sin(k_s z + \gamma) & z > d \end{cases}$$



Bulk—longitudinal waves:
the bulk particles oscillate or vibrate in the same axis of
the wave propagation

(Introduction to Acoustic Wave Resonators,
H. Campanella, ARTECH HOUSE, 2010)

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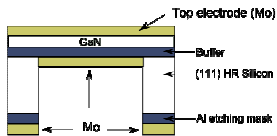
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FBAR Fabrication Using Bulk Micromachining



GaN membrane FBAR structures



- Mo deposition and **lift-off** techniques to define the FBAR structures on the top.
- Second mask used to cover Mo with a gold layer (500nm)
- **Backside lapping** of the wafer to a thickness of about 150μm.
- Al deposition on the bottom (as mask during RIE of silicon)
- **Backside patterning** for the membrane formation.
- Backside RIE of silicon down to the 2.2μm thin GaN layer using SF₆ plasma.
- Sputtering of thin Mo layer for backside floating electrode formation

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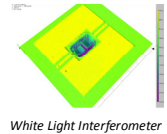
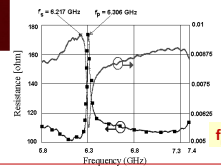
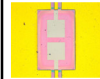
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FBAR Fabrication Using Bulk Micromachining

6.3 GHz resonance on a GaN FBAR obtained by micromachining of GaN/Si

-540nm (340 nm (GaN) +200nm (buffer)) thin membrane supported FBAR structure based on GaN micromachining

-50nm thin Mo metallization GaN/Si wafers from NTT AT Japan



White Light Interferometer

A. Müller, D. Neculoiu, G. Konstantinidis et al. "6.3 GHz Film Bulk Acoustic Resonator Structures Based on a Gallium Nitride/Silicon Thin Membrane"

IEEE Electron Devices Letters, vol 30, no 8, August 2009, pp 799-801

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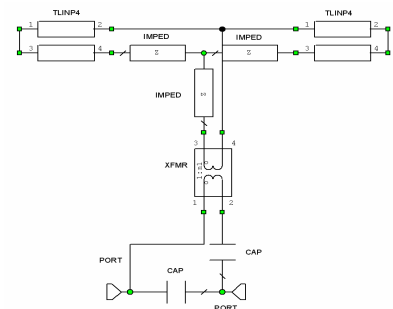
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Analytical Modeling

Mason's circuit



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Analytical Modeling

Mason's model

$$Z_m = \frac{1}{j\omega C_0} \left[1 - \frac{k_T^2}{\gamma} \frac{(z_1 + z_2) \sin \gamma + 2j(1 - \cos \gamma)}{(z_1 + z_2) \cos \gamma + j(1 + z_1 z_2) \sin \gamma} \right]$$

$$C_0 = \epsilon_{33} S / l$$

$$\gamma = \omega l / V$$

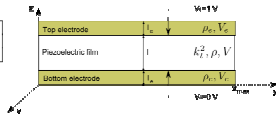
$$V = \sqrt{\frac{E}{\rho}}$$

$$z_1, z_2$$
 acoustic load impedances on both sides of the GaN layer normalized to $Z_0 = S\rho V$

$$k_T^2 = \frac{\frac{e_{33}^2}{\epsilon_{33}}}{1 + \frac{e_{33}^2}{\epsilon_{33}}}$$

electromechanical coupling coefficient of the piezoelectric layer

Simulated FBAR geometry



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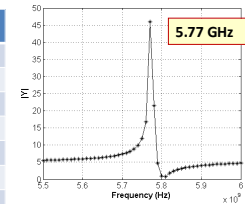
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Analytical Modeling

The parameters of the composite FBAR structure used in simulations

	Electrodes (Mo)	Piezoelectric film (GaN)
ρ [kg/m ³]	10220	6150
V [m/s]	67	5844
l [μ m]	0.06	0.5
$\mathcal{E}_T$	-	9
e_{31} [Cm ⁻²]	-	-0.36
e_{33} [Cm ⁻²]	-	1
e_{15} [Cm ⁻²]	-	-0.3
C_{33} [GPa]	330	398



Simulated input admittance of FBAR versus frequency (MATLAB)

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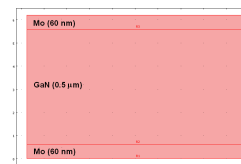
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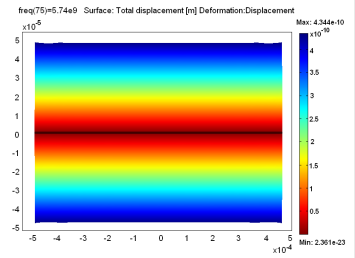
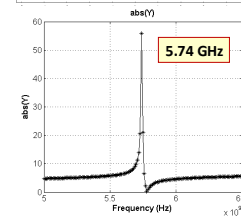
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FEM Modeling



COMSOL model MEMS module (Piezo Plane Strain)

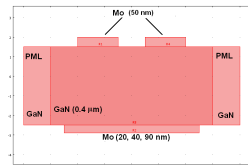


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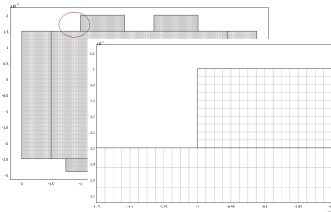
FEM Modeling



COMSOL model - MEMS module
(Piezo Plane Strain + Plane Strain)

Mesh 13440 elements
153 176 dofs

	Width [m]	Height [m]
Bottom electrode	2.5e-4	40e-9
		20e-9
		90e-9
Top electrode	7.5e-5	50e-9
Piezolayer	3e-4	0.4e-6



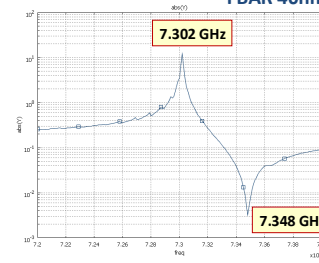
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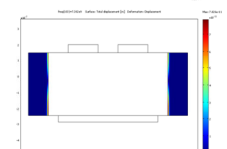
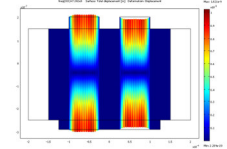
FEM Modeling

FBAR 40nm Results



Simulated input admittance of
FBAR versus frequency

Total displacement



Total displacement on the PML

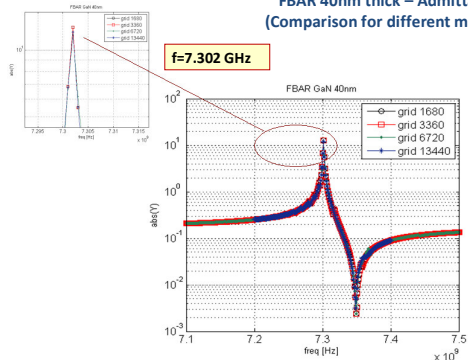
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FEM Modeling

FBAR 40nm thick – Admittance
(Comparison for different meshes)



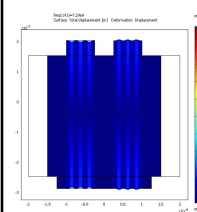
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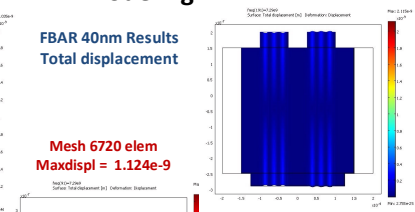
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FEM Modeling

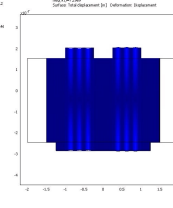
FBAR 40nm Results
Total displacement



Mesh 3360 elem
Maxdispl = 1.02e-9



Mesh 6720 elem
Maxdispl = 1.124e-9



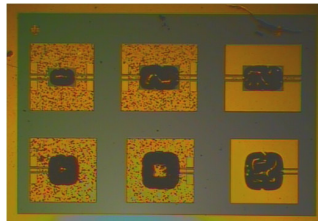
Mesh 13440 elem
Maxdispl = 1.12e-9

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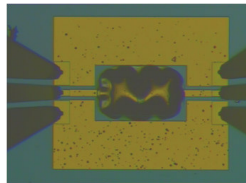
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FEM Modeling



Fabricated chip with series and parallel connected FBARs

Photo of two FBARs in series configuration

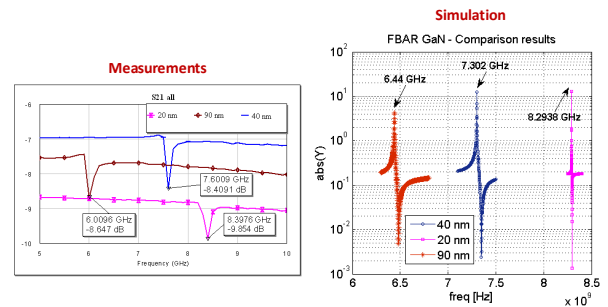


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FEM Modeling FBAR Comparison Results



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Conclusions

- **Modeling approaches for FBAR structure:**
 - ✓ Analytical model for film bulk acoustic resonator (physics of the resonator, equivalent circuit model)
 - ✓ Finite Element Models based on numerical methods and domain coupling analysis
- **Good agreement between simulation and experimental data previously reported**
- **Next step: 3D accurate model analysis**

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Feedback/Questions

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Thank you for attention!

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